OPTIMIZATION OF MULTIPLIER IN REVERSIBLE LOGIC

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ABSTRACT

Reversible logic is leading area in power consumption. Based on its application, its emerging trend in power consumption. In ideal situations, reversible circuit yield nil power. Different methods of multiplier optimized in this paper. Like quantum computers, multipliers required to develop computational units. In the paper, two different methods of multiplier developed with very large bit width. Based on partial products hierarchical method is developed. Another method is Karatsuba's algorithm developed based on divide and conquers method. Finally, we compare the results of both two methods. The projected reversible multipliers are enhanced in terms of quantum cost, number of constant inputs, number of garbage outputs and complexity in hardware. In nanotechnology applications this multiplier can be used to construct complex system.

KEYWORDS

Karatsuba's algorithm, Hierarchical method, FFT, Reversible gates.

1. INTRODUCTION

One of the foremost issues in VLSI design is reducing power. In early days, due to information loss irreversible logic leads to power dissipation (Zhirnov *et al.*, 2003). When one bit information lost, it leads to dissipate at least KTln2 joules of energy, where $K=1.380650 \times 10-23 \text{ m}2\text{kg}-2\text{K}-1$ (joules Kelvin-1) is the Boltzmann's constant and T is the temperature (Zhirnov *et al.*, 2003). Reversible logic blocks do not lose information because it has internally zero power dissipation. To eliminate KTln2 joules of energy dissipation, circuit must be developed with reversible logic gates.

In the subsequent, based on the reversible logic multipliers are synthesized. Initially multipliers synthesized based on Toffoli circuit (Karatsuba & Ofman, 1963; Islam *et al.*, 2009). Also, multiplier can be synthesized based on different reversible gates. In addition to that, multipliers have been proposed with large bit-width.

This paper proposes a multiplier in reversible logic with large bit-width. Two methods are projected, the foremost method is hierarchical method and another one is Karatsuba method applied to the application of FFT.

Respite of the paper systematized as follows. Reversible logic gates ideas and necessity described in Segment II. In Segment III, basics of reversible functions and circuits are discussed. Segment IV provides algorithm of proposed work. Segment V provides the design of Hierarchical and Karatsuba method. Segment VI determines the conclusion.

2. REVERSIBLE LOGIC GATES

In segment II, basic reversible logic gates are described which is being used in the design. Figure 1 displays a 3*3 Toffoli gate. The input path is I (A, B and C) and the output path is O (P, Q, and R). The outputs of the gate are well defined by P=A, Q=B, $R=AB\bigoplus C$.

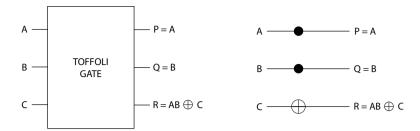
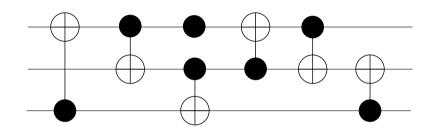


Figure 1. Toffoli Gate. Source: own elaboration.



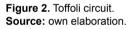


Figure 2 shows a Toffoli circuit with 6 gates, quantum cost value 10, cost of transistor 56 and 3 circuit lines respectively. Black circle \bullet is denoted as control lines and the cross line \oplus denoted as target lines.

3. BASICS OF REVERSIBLE LOGIC

Demarcation 1: A reversible function $f : \mathbb{B}^n \to \mathbb{B}^m$ is defined as if (1) its amount of outputs is same as the amount of inputs and (2) it plots every input design to a distinctive output.

Reversible function is required to be embed in irreversible to reduce data loss, which requires circuit lines to produce constant inputs and garbage outputs. The least count of circuit lines is added, and it determined by number of existences of the most numerous output pattern (Toffoli, 1980). Reversible functions realized based on restrictions as if fan-out and feedback are not indorsed (Wille & Drechsler, 2009).

Demarcation 2: Inputs in reversible circuit G $X = \{x_i \mid 1 \le i \le n\}$ determined by reversible gates gi cascaded, i.e. $G = \prod_{i=1}^{d} g_i$ gi where the number of gate denoted as d. In general, different reversible gates are there. In which most used gate is Toffoli because it is universal, multiple gate control.

Demarcation 3: Let the domain function denoted as $X := \{x_i \mid 1 \le i \le n\}$. g(C; t) is the form of universal multiple control Toffoli gate, where $C = \{x_{ji} \mid 1 \le i \le k\} \subset X$ control lines set denoted as X and with t 62 C the target line denoted as t = xl. The target line is inverted when control lines set assigned as one, that is., When control line is empty the input vector of the gate has been charted to $T_{i=1}^{i_1}x_{i_1}x_i \oplus 1$, $T_{i=1+1}^nx_i$, then the target is inverted. In the subsequent, Toffoli gates are referred as multiple control gate. Toffoli gate also called as Not gate when it has zero control lines.

4. PROPOSED REVERSIBLE MULTIPLIER

4.1. HIERARCHICAL METHOD

In this segment, multiplier synthesized based on hierarchical method by means of controlled increasers. To calculate partial products, multiply two factors $a = \sum_{i=0}^{n-1} a_i \cdot 2^i$ and $b = \sum_{i=0}^{n-1} b_i \cdot 2^i$ then add the products together, i.e. $a \cdot b = \sum_{i=0}^{n-1} (a_i \sum_{i=0}^{n-1} b_j \cdot 2^i) \cdot 2^i$. When the bit ai assigned as one then the particular bit of bj multiplied by particular rule of 2 is additional to the product. This can be apprehended by controlled functions. Therefore, by using hierarchical method number of multiple n-bit apprehends multiplier by distinct n-bit controlled increasers. The factors of multiple bit multiplication are $a = \prod_{i=0}^{n-1} a_i$ and $b = \prod_{i=0}^{n-1} b_i$ as well as the product $c = \prod_{i=0}^{2n-1} c_i$ here, the i-th controlled increaser is controlled by ai. It conditionally adds the value of b to $\prod_{j=i}^{n-1+i} C_i$ i.e. to the n bits of the product c beginning from the i-th bit. To modify the j-th bit of the product, lower bit does not consider until the j-th controlled increaser. Therefore, by a bit shifter, we can realize the controlled increasers without considering lower bits. The n + i-th position of the product carries carry, which further not used and, thus, carries zero.

1 for i = 0 to n - 12 { 3 $T_{j=i}^{n+i} c_i + \stackrel{a_i}{=} T_{i=0}^{n-1} b_i$ 4 }

Example 1: Consider a 3-bit factors based on hierarchical a = a2a1a0 and b = b2b1b0 along with the product c = 5c4c3c 2c1c0. To realize this kind of multiplication we required 3 controlled increasers. a0 is control the initial controlled increaser. In addition to that it temporarily added 2^{nd} factor b to last 3 LSB of product c2c1c0. Carry writes over to third MSB bit c3. a1 is control the 2^{nd} controlled increaser. In addition to that it temporarily added 2^{nd} factor b to c3c2c1. Carry writes over to second MSB bit c4. a2 controls the 3rd controlled increaser. In addition to that it temporarily added 2^{nd} factor b with product c, i.e. to c3c2c1. Carry writes over to second MSB bit c4. bit controlled increaser. In addition to that it temporarily added 2^{nd} factor b with product c, i.e. to c3c2c1. Carry writes over to second MSB bit c4.

4.2. KARATSUBA METHOD

In this segment Karatsuba's algorithm, based multiplier designed with divide and conquer method. Based on this method multiplication realized by multiplying two factors with smaller bit-width and additionally perform some less expensive operations. Consider an n-bit multiplication with m = 2.k, Both multiple factors (example. $a = \sum_{i=0}^{2 \cdot k \cdot 1} a_i \cdot 2^i$ are split as an upper part $\overline{a} := \sum_{i=k}^{2 \cdot k \cdot 1} a_i \cdot 2^{i \cdot k}$ and a lower part $\underline{a} := \sum_{i=0}^{k \cdot 1} a_i \cdot 2^i$ such that $a = \overline{a} \cdot 2^k + a$. With this demonstration the subsequent equations are:

$$\begin{split} a \cdot b &= \left(\overline{a} \cdot 2^k + \underline{a}\right) \cdot \left(\overline{b} \cdot 2^k + \underline{b}\right) \\ &= \overline{a} \cdot \overline{b} \cdot 2^{2 \cdot k} + \left(\underline{a} \cdot \overline{b} + \overline{a} \cdot \underline{b}\right) \cdot 2^k + \underline{a} \cdot \underline{b} \\ &= \overline{a} \cdot \overline{b} \cdot 2^{2 \cdot k} + \left(\underline{a} \cdot \overline{b} + \overline{a} \cdot \underline{b} + \overline{a} \cdot \overline{b} + \underline{a} \cdot \underline{b} - \overline{a} \cdot \overline{b} - \underline{a} \cdot \underline{b}\right) \cdot 2^k + \underline{a} \cdot \underline{b} \\ &= \overline{a} \cdot \overline{b} \cdot 2^{2 \cdot k} + \left(\underline{a} \cdot (\underline{b} + \overline{b}) + \overline{a} \cdot (\underline{b} + \overline{b}) - \overline{a} \cdot \overline{b} - \underline{a} \cdot \underline{b}\right) \cdot 2^k + \underline{a} \cdot \underline{b} \\ &= \overline{a} \cdot \overline{b} \cdot 2^{2 \cdot k} + \left(\underline{a} \cdot (\underline{b} + \overline{b}) + \overline{a} \cdot (\underline{b} + \overline{b}) - \overline{a} \cdot \overline{b} - \underline{a} \cdot \underline{b}\right) \cdot 2^k + \underline{a} \cdot \underline{b} \\ &= \overline{a} \cdot \overline{b} \cdot 2^{2 \cdot k} + \left((\underline{a} + \overline{a}) \cdot (\underline{b} + \overline{b}) - \overline{a} \cdot \overline{b} - \underline{a} \cdot \underline{b}\right) \cdot 2^k + \underline{a} \cdot \underline{b} \end{split}$$

These expressions indicate that 3 k-bit multiplications, bit shits and subtractions realized from a (2.k)-bit multiplication. In addition to that, circuit lines are necessary for reversible logic. For small bit width Karatsuba method is not applicable, the variable turning Point

signifies the bit-width lower. The Karatsuba method-based multiplication is used afar the turning Point.

Example 2:

Let us deliberate eight-bit Karatsuba multiplication and turning point t = 6. The turning point t is less than 8 bit and even the conditional lines 1 and 4 does not hold and computed K as four (line 7). Then, in line eight, new variables d, e, f are initialized which leads to 20 garbage lines i.e 4.4+4 = 20. Then, the 2 minor multiplications c= c7c6c5c4c3c2c1c0 =a3a2a1a0_b3b2b1b0 = $\underline{a} \cdot \underline{b}$ (line 9) and $\overline{c} = \overline{a} \cdot \overline{b}$ (line 10) are performed, respectively. In previous segment, same method realized using hierarchical method. Then, the result of the multiplications directly consigned to the product of the bits. To modify the product of the sums this result must be used that calculated next. These sums are d=d4d3d2d1d0= a7a6a5a4+a3a2a1a0 = $\overline{a} + \underline{a}$ (line 11) and $e = \overline{b} + \underline{b}$ (line 12). To perform this operation, copy the 1st sum to the target until it uninitialized and increase the function by adding 2nd sum. To get the 3rd sub product, result of this 1st and 2nd sums are multiplied h = d.e (line 13).

```
1 if (n < turningPoint)
         c = MULT_H(a,b)
2
3
4 if (n \% 2 = 1)
               init a_n, b_n, c_{2 \cdot n+1 \text{ with } 0}
5
6
7 k := \lfloor \frac{n}{2} \rfloor
8 init d, e^{2}(k+1 \text{ bits}), f(2 \cdot k+2 \text{ bits}) with 0
9 \overline{c} = MULT_K(\overline{a}, \overline{b})
10 c = MULT_K(a, b)
11 d = \overline{a} + a
12 e = \overline{b} + b
13 h = MULT_K(d, e)
14 h = \overline{c}
15 h - = c
16 \mathsf{T}_{i=k}^{3\cdot \overline{k}+3}
               c_i + = h
```

Multiplication performed by the hierarchical approach when turning point is greater than the bit.

5. DESIGN OF REVERSIBLE MULTIPLIER

5.1. HIERARCHICAL METHOD

| Messages | | |
|---------------------|---|---|
| /testmult 16/dk | 0 | |
| /testmult 16/resetb | 1 | |
| /testmult16/start | 0 | η |
| + 🔶 /testmult16/a | 000000000001011 | 000000000000000000000000000000000000000 |
| + 🔶 /testmult16/b | 000000010000001 | 0000000010000001 |
| + 🔶 /testmult16/y | 000000000000000000000000000000000000000 | 0000)0000 00000000000000000000000 |
| /testmult 16/done | St1 | |
| /testmult 16/num_er | 0 | 0 |

5.2. KARATSUBA METHOD

For example, the same values can be multiplied using karatsuba method representing with 16bits.

| | Messages | | | | |
|-----|--|---|------|---|--|
| | /testmult 16/dk /testmult 16/resetb | 0 | | wwww | น้ำแนนแนนแนนแนนแนนแนนแนนแนนแนนแนนแนนแนนแนน |
| | /testmult 16/start | 0 | Ľ | 1 | |
| | /testmult16/a | 000000000001011 | | 000000000000000000000000000000000000000 | 001011 |
| ± 🔶 | /testmult16/b | 0000000010000001 | | 000000010 | 000001 |
| • | /testmult16/y | 000000000000000000000000000000000000000 | 0000 | .)0000)0 | 000000000000000000000000000000000000000 |
| | /testmult16/done /testmult16/num_er | St1 0 | 0 | | |

6. CONCLUSIONS

This paper proposes a multiplier with very large bit-width using reversible logic. The Hierarchical method only applicable for small bit-width but the Karatsuba method applicable for large bit-width. The results are compared using synthesized result of device utilization of the two methods. The Karatsuba method is better than the Hierarchical method because the device utilization was more in the case of Hierarchical method that is mention in Table 1. The projected reversible multipliers are enhanced in terms of quantum cost, number of constant inputs, number of garbage outputs and complexity in hardware. In nanotechnology applications this multiplier can be used to construct complex system.

This result is taken by 16bit multiplication of two binary values 000000000001011 and 0000000010000001.

Table 1. Comparison of proposed systems.

| | NO. OF FLIP FLOPS | NO. OF 4INPUT LUTS | NO. OF IOBS |
|---------------------|----------------------|-----------------------|-------------|
| Hierarchical method | 81 | 122 | 68 |
| Karatsuba method | 36 | 72 | 64 |

Source: own elaboration.

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